L295

DUAL SWITCH-MODE SOLENOID DRIVER

- HIGH CURRENT CAPABILITY (up to 2.5A per channel)
- HIGH VOLTAGE OPERATION (up to 46V for power stage)
- HIGH EFFICIENCY SWITCHMODE OPERATION
- REGULATED OUTPUT CURRENT (adjustable)
- FEW EXTERNAL COMPONENTS
- SEPARATE LOGIC SUPPLY
- THERMAL PROTECTION

MULTIWATT15
ORDERING NUMBER: L295

DESCRIPTION

The L295 is a monolithic integrated circuit in a 15-lead Multiwatt® package; it incorporates all the functions for direct interfacing between digital circuitry and inductive loads.
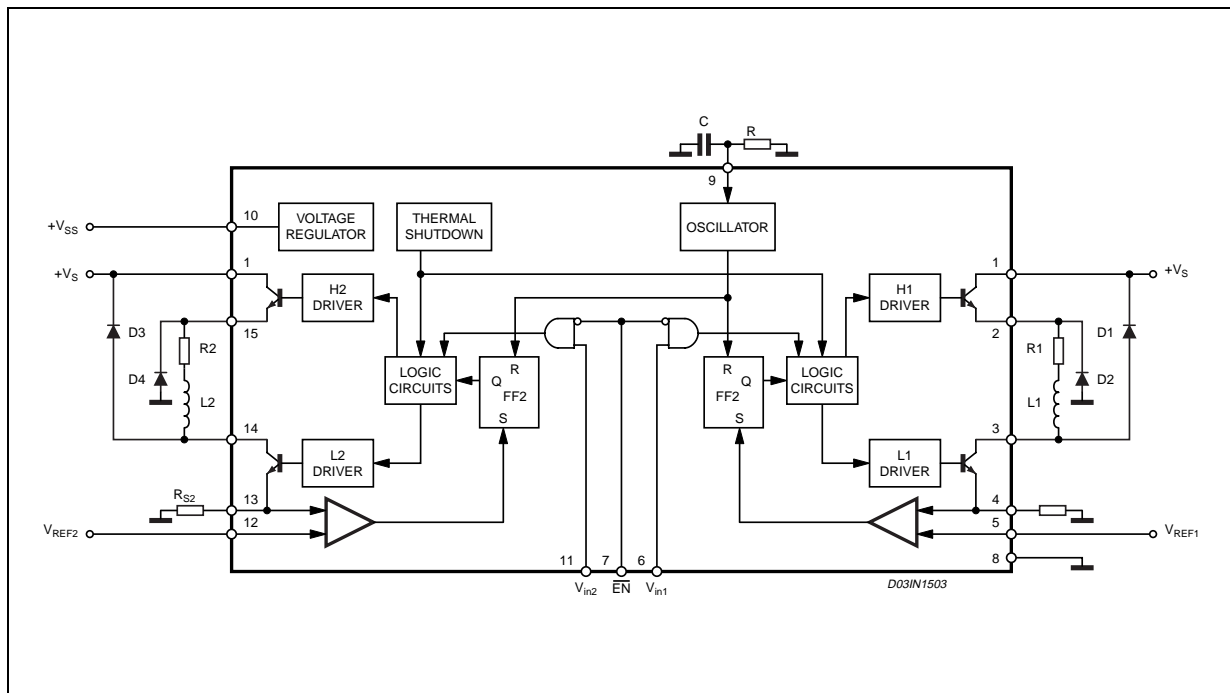
The L295 is designed to accept standard microprocessor logic levels at the inputs and can drive 2 solenoids. The output current is completely controlled by means of a switch-ing technique allowing very efficient operation.

Furthermore, it includes an enable input and dual supplies (for interfacing with peripherals running at a higher voltage than the logic).

The L295 is particularly suitable for applications such as hammer driving in matrix printers, step motor driving and electromagnet controllers.
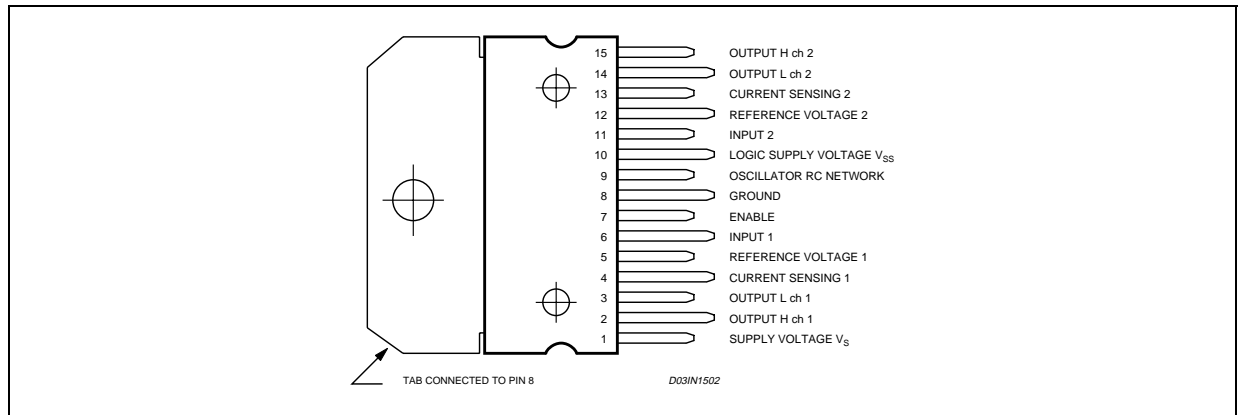
BLOCK DIAGRAM

D03IN1503

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_S	Supply voltage	50	V
V_{SS}	Logic supply voltage	12	V
V_{EN}, V_i	Enable and input voltage	7	V
V_{ref}	Reference voltage	7	V
I_o	Peak output current (each channel) - non repetitive (t = 100 µsec) - repetitive (80% on - 20% off; T_{on} = 10ms) - DC operation	 3 2.5 2	 A A A
P_{tot}	Total power dissipation (at T_{case} = 75 ″C)	25	W
T_{stg}, T_j	Storage and junction temperature	- 40 to 150	°C

CONNECTION DIAGRAM

15	OUTPUT H ch 2
14	OUTPUT L ch 2
13	CURRENT SENSING 2
12	REFERENCE VOLTAGE 2
11	INPUT 2
10	LOGIC SUPPLY VOLTAGE V_{SS}
9	OSCILLATOR RC NETWORK
8	GROUND
7	ENABLE
6	INPUT 1
5	REFERENCE VOLTAGE 1
4	CURRENT SENSING 1
3	OUTPUT L ch 1
2	OUTPUT H ch 1
1	SUPPLY VOLTAGE V_S

TAB CONNECTED TO PIN 8   D03IN1502

THERMAL DATA

Symbol	Parameter	Value	Unit
$R_{th-j-case}$	Thermal resistance junction-case	max 3	°C/W
$R_{th-j-amb}$	Thermal resistance junction-ambient	max 35	°C/W

ELECTRICAL CHARACTERISTCS (Refer to the application circuit, Vss = 5V, Vs = 36V; Tj = 25°C; L = Low; H = High; unless otherwise specified)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V_S	Supply Voltage		12		46	V
V_{SS}	Logic Supply Voltage		4.75		10	V
I_d	Quiescent Drain Current (from V_{SS})	V_S = 46V; V_{i1} = V_{i2} = V_{EN} = L			4	mA
I_{ss}	Quiescent Drain Current (from V_S)	V_{SS} = 10V			46	mA
V_{i1}, V_{i2}	Input Voltage	Low High	-0.3 2.2		0.8 7	V V

ELECTRICAL CHARACTERISTCS (continued)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
V_{EN}	Enable Input Voltage	Low High	-0.3 2.2		0.8 7	V V
I_{i1}, I_{i2}	Input Current	$V_{i1} = V_{i2} = L$ $V_{i1} = V_{i2} = H$			-100 10	µA µA
I_{EN}	Enable Input Current	$V_{EN} = L$ $V_{EN} = H$			-100 10	µA µA
V_{ref1}, V_{ref2}	Input Reference Voltage		0.2		2	V
I_{ref1}, I_{ref2}	Input Reference Voltage				-5	µA
F_{osc}	Oscillator Frequency	C = 3.9 nF; R = 9.1 KΩ		25	46	KHz
$\dfrac{I_p}{V_{ref}}$	Transconductance (each ch.)	$V_{ref} = 1V$	1.9	2	2.1	A/V
V_{ref1}, V_{ref2}	Input Reference Voltage		0.2		2	V
V_{ref1}, V_{ref2}	Input Reference Voltage		0.2		2	V
V_{drop}	Total Output Voltage (each channel) (*)	$I_o = 2 A$		2.8	3.6	V
V_{sens1}, V_{sens2}	External sensing resistors voltage drop				2	V

(*) $V_{drop} = V_{CEsat\ Q1} + V_{CEsat\ Q2}$.

APPLICATION CIRCUIT
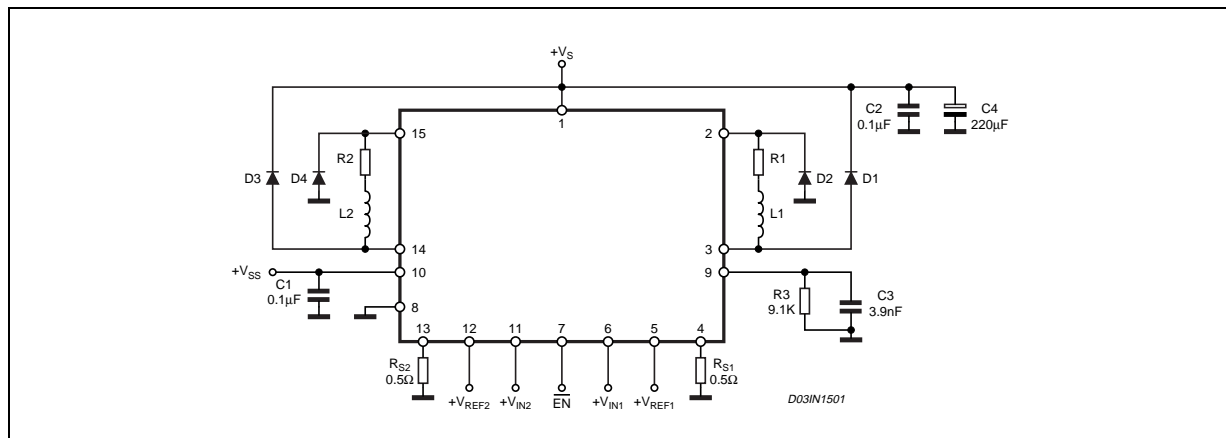
D2, D4 = 2A High speed diodes trr ≤ 200 ns
D1, D3 = 1A High speed diodes trr ≤ 200 ns

R1 = R2 = 2W
L1 = L2 = 5 mH

FUNCTIONAL DESCRIPTION

The L295 incorporates two indipendent driver channals with separate inputs and outputs, each capable of driving an inductive load (see block diagram). The device is controlled by three micriprocessor compatible digital inputs and two analog inputs.

These inputs are:
– **EN** chip enable (digital input, active low), enables both channels when in the low state.

– **V_{in1}, V_{in2}** channel inputs (digital inputs, active high), enable each channel independently.
  A channel is actived when both EN and the appropriate channel input are active.

– **V_{ref1}, V_{ref2}** referce voltages (analog inputs), used to program the peak load currents. Peak load current is proportional to V_{ref}.

Since the two channels are identical, only channel one will be described. The following description applies also the channel two, replacing FF2 for FF1, V_{ref} for V_{ref1} etc.

When the channel is avtivated by low level on the EN input and a high level on the channel input, V_{in2}, the output transistors Q1 and Q2 switch on and current flows in the load according to the exponential law:

$$I = \frac{V}{R1}\left(1 - e^{\frac{-R1t}{L1}}\right)$$

where:

R1 and R2 are the resistance and inductance of the load and V is the voltage available on the load (V_s - V_{drop} - V_{sense}).

The current increases until the voltage on the external sensing resistor, R_{S1}, reaches the reference voltage, V_{ref1}. This peak current, I_{p1}, is given by:

$$I_{p1} = \frac{V_{ref1}}{R_{S1}}$$

At this point the comparator output, Vomp1, sete the RS flip-flop, FF1, that turns off the output transistor, Q1. The load current flowing through D2, Q2, R_{S1}, decreases according to the law:

$$I = \left(\frac{V_A}{R_1} + I_{p1}\right) e^{\frac{-R1t}{L1}} - \frac{V_A}{R1}$$

where $V_A = V_{CEsat\,Q2} + V_{sense} + V_{D2}$

If the oscillator pin (9) is connected to ground the load current falls to zero as shown in fig. 1.

At this time t_2 the channel 1 is disabled, by taking the inputs V_{in1} low and/or EN high, and the output transistor Q2 is turned off. The load current flows through D2 and D1 according to the law:

$$I = \left(\frac{V_B}{R_1} + I_{T2}\right) e^{\frac{-R1t}{L1}} - \frac{V_B}{R1}$$

where $V_B = V_S + V_{D1} + V_{D2}$

I_{T2} = current value at the time t_2.

Fig. 2 in shows the current waveform obtained with an RC network connected between pin 9 and ground. From to t_1 the current increases as in fig.1. A difference exists at the time t_2 because the current starts to increase again. At this time a pulse is produced by the oscillator circuit that resets the flip. flop, FF1, and switches on the outout transistor, Q1. The current increases until the drop on the sensing resistor RS1 is equal to V_{ref1} (t_3) and the cycle repeats.

The switching frequency depends on the value R and C, as shown in fig. 4 and must be chosen in the range 10 to 30 KHz. It is possible with external hardware to change the reference voltage V_{ref} in order to obtain a high peak current I_p and a lower holding current I_h (see fig. 3).

The L295 is provided with a thermal protection that switches off all the output transistors when the junction temperature exceeds 150°C. The presence of a hysteresis circuit makes the IC work again aftera fall of

the junction temperature of about 20°C.

The analog input pins (V_{ref1}, V_{ref2}) can be left open or connected to V_{ss}; in this case the circuit works with an internal reference voltage of about 2.5V and the peak current in the load is fixed only by the value of R_S:

$$I_P = \frac{2.5}{R_S}$$

SIGNAL WAVEFORMS

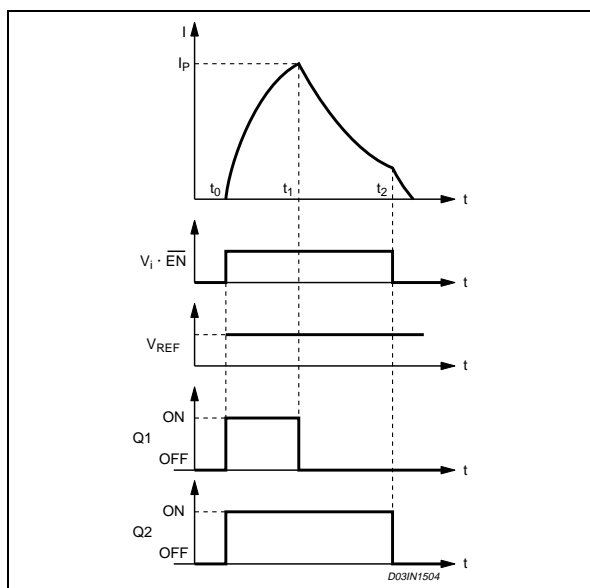**Figure 1. Load current waveform with pin 9 connected to GND.**

Figure 2. Load current waveform with external R-C network connected between pin 9 and ground.
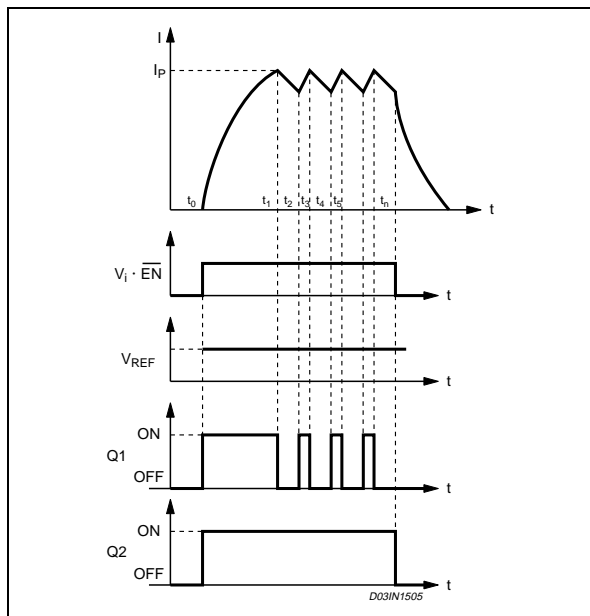
Figure 3. With V_{ref} changed by hardware.
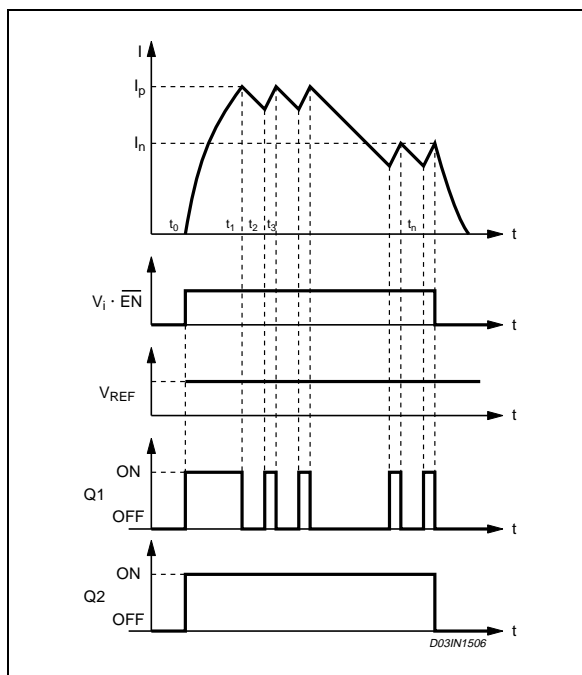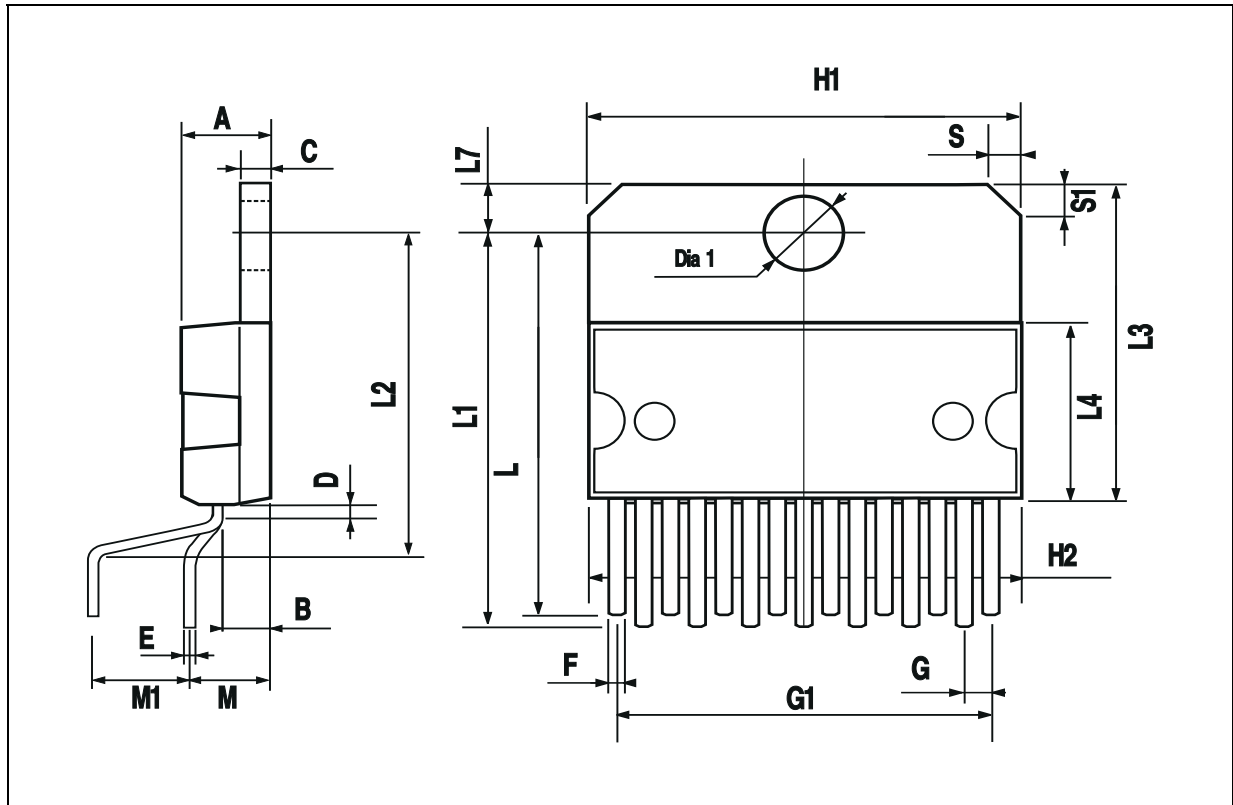

Figure 4. Switching frequency vs. values of R and C.
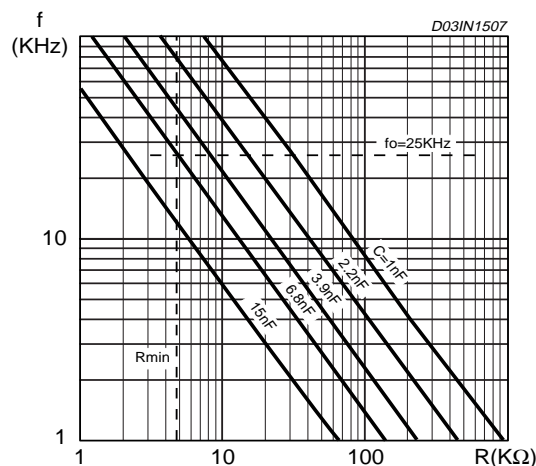
DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			5			0.197
B			2.65			0.104
C			1.6			0.063
D		1			0.039	
E	0.49		0.55	0.019		0.022
F	0.66		0.75	0.026		0.030
G	1.02	1.27	1.52	0.040	0.050	0.060
G1	17.53	17.78	18.03	0.690	0.700	0.710
H1	19.6			0.772		
H2			20.2			0.795
L	21.9	22.2	22.5	0.862	0.874	0.886
L1	21.7	22.1	22.5	0.854	0.870	0.886
L2	17.65		18.1	0.695		0.713
L3	17.25	17.5	17.75	0.679	0.689	0.699
L4	10.3	10.7	10.9	0.406	0.421	0.429
L7	2.65		2.9	0.104		0.114
M	4.25	4.55	4.85	0.167	0.179	0.191
M1	4.63	5.08	5.53	0.182	0.200	0.218
S	1.9		2.6	0.075		0.102
S1	1.9		2.6	0.075		0.102
Dia1	3.65		3.85	0.144		0.152

OUTLINE AND MECHANICAL DATA

Multiwatt15 V

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